Setup of Cryogenic Front-End Electronic Systems for Germanium Detectors Read-Out

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Front-end electronic devices for the read-out of ionizing radiation sensors must operate in many cases at cryogenic temperatures. Sometimes the front-end circuit is divided into a cold part operated at cryogenic temperatures and a warm part operated at room temperature outside the experimental setup, the detector system requirements or the apparatus geometry. In this latter cases the front-end circuitry has to operate in its entirety at cryogenic temperature. In this work, carried on in the framework of the GERDA experiment (GERmanium Detector Array), we focus in particular on front-end read-out systems for High-Purity Germanium (HPGe) detectors, which are usually operated at liquid nitrogen (LN) temperature. We study the strong effects that the changed characteristics of the electronic active and passive devices have on the charge preamplifier design when operated in LN, while taking into account the particularly challenging requirements that the circuit has to meet: radio-purity, physical reliability under thermal cycling, good noise performance (0.1-0.2% resolutions) and fast rise time (20 ns) needed for pulse shape analysis applications. We discuss in particular the effects that changes of JFET and MOSFET transconductance have on noise and bandwidth performance of the fort-end circuit. We also discuss my have on the preamplifier response when operated in LN.

Requirements for cryogenic charge sensitive preamplifiers for High-Purity Germanium detectors

- Full functionality at cryogenic temperature (77K in liquid nitrogen, 86K in liquid argon)
- Physical reliability under thermal cycling
- High degree of radio-purity (when inserted in ultra-low background radiation detection systems)
- Low noise (gamma spectroscopy grade: 0.1-0.2 % the intrinsic resolution of large-volume HPGe detectors is ~1.6 keV @ 1.3 MeV)
- Excellent stability of the gain (loop gain of the order of 10³)
- Wide bandwidth: rise time of ~ 20 ns (in order to apply the pulse shape analysis algorithms)
- Large dynamic range: at least ~10³ (10 keV -10 MeV)
- Low power consumption (~20-40mW per read-out channel)
- Output stage able to drive long 50Ω-terminated coaxial cables

V_10 = 2 V

0.75 V 0.5 V

Room temperature (300K) vs. cryogenic temperature (77K) behavior of active and passive devices

JFET BF862 rated at T = 300 K

V___(V)

V...= 2 \

2

Measured transconductance of the BF862 JFET manufactured by Philips: a drop of a factor 5 is measured at 77K

Silicon <u>Bipolar Junction Transistors (BJTs</u>) cannot operate at 77K, because their performances are completely spoiled by freeze-out phenomena.

For a <u>MOSFET</u> of a given geometry, transconductance is proportional to the drift velocity in the channel. At low channel fields (long channel, small drain voltage) transconductance is proportional to the low-field carrier mobility. Mobility increases while decreasing the temperature from 300K to 77K by a factor of 4-6, because of the reduced carrier scattering due to lattice vibration. This increase in mobility is also observed on long channels in the saturation mode, where it can be assumed that the drift velocity in the channel is proportional to the longitudinal field

The <u>Silicon JFET</u> transconductance increases while temperature decreases down to 120K. Then the transconductance decreases again, owing to the increased scattering due to impurities in the lattice. At about 40K the JFET stops working because of carrier freeze-out. As the transconductance is inversely proportional to the thermal noise, the operating point at 120K is the optimal one. The choice of using a JFET as input transistor at 77K, even though thermal noise is not optimized, can be dictated by the extremely low level of 1/f-like noise that JFETs exhibit, which can be particularly important for spectroscopic applications

> JFET BF862 wated at T = 77 K







mperature dependence of channel mobility for both WLF-karek, DEF-karek, NCS-Fries, OCE-HOLDING WAS ctrons (NMOS) and holes (PMOS), as measured in linear mode on a 0.5 µm CMOS technology. VI. Loank, D. Er-karek, NCS-Fries, OCE-HOLDING III (Linear mode on a 0.5 µm CMOS technology. vol. 15, pp. 397-404, 1992





Single-channel preamplifier scheme and ASIC layout. The ASIC has been realized in a mature CMOS CXZ 0.8 µm 5V technology provided by Austria Micro Systems. The area occupancy of the ASIC is as little as 366x275 µm² excluding the bondine pads.



Noise and bandwidth performance of the realized preamplifier at 300K and 77K

V___ (V)



Disturbing effects of the cryogenic setup on the preamplifier performances

- ✓ The spoiled performance of the high-value ceramic capacitances yields disturbing effects on the shape of the preamplifier response: when large and fast signals have to be provided on a low output load (like a terminated coaxial cable), the circuit has to deliver a considerable power in the fastest possible time and this is achieved by means of the charge stored on the high-value capacitances used for power supplies filtering → tantalum capacitors, maintaining their constant value at 77K, can be used at the expenses of a high dissipation factor.
- ✓ Use of thin cables is mandatory to achieve a high level of flexibility and maintain a maximum level of radio-purity, at the expenses of a high cable resistance up to 0.8-0.9 ohm/m. Since long cables (10-12m) may be needed in a large-dimensions cryogenic setup (like that of GERDA) to connect the circuit with the outside, a series resistance of 10-12 ohm may be present along power supplies cables.
- ✓ <u>Power-supply bounce</u> is encountered due to the high resistance cables and its effect gets enhanced by the lack of a needed filtering capacitance. This can also yield <u>cross-talk effects in multi-channel configurations</u> → a particular care must be used in separating power supplies of the input and output stages, so as to separate the main gain stages from the stages where power has to be delivered.
- Imand 10m RG58 cable The cable resistance on the signal. Use of long and thin cable is often mandatory in cryogenic setup.





Output signal of two different preamplifier channels operated at 77K and connected to the outside through 12-m RG168 cables: when one common power supply is used, and a considerable power has to be delivered, the output exponential signal shape is distorted by the <u>bounce effect on the common power supply</u>. An appropriate separation of the input/output stages power supplies is mandatory to eliminate the distortion.